Lab 1.2
Joystick Interface

Lab 1.0 + 1.1 – PWM Software/Hardware Design (recap)
The previous labs in the 1.x series put you through the following progression:

- Lab 1.0 – You learnt some theory behind how one interacts with a hardware peripheral, then went on to practice what you learnt by means of a programming exercise. The practical task involved writing the control software in C for a black-box PWM generator connected to a pan-tilt module.
- Lab 1.1 – The control software written in lab 1.0 was left untouched, but you replaced the black-box PWM generator with a custom hardware design you wrote yourself in VHDL. You also learnt how to write a VHDL testbench so you can test the hardware components you will be developing throughout the course.

Lab 1.2 – Joystick Interface
Theory
Introduction
The previous labs were centered around output devices, so it’s now time to switch gears and explore input devices. We use the scenario below as motivation for this lab:

Until this point, the control software was issuing the movement pattern of the servomotors by means of a hard-coded algorithm in the C source code. We now want to allow the user to provide the position of the servomotors by means of an input device, a PlayStation 2 joystick. We follow the steps below to this end:

1. Remove the hard-coded movement pattern from the control software and push the burden of providing the expected position of the servomotors to the user. This can be done by adding a new layer of indirection by means of an input device.
2. The user interactively manipulates the input device to provide his/her desired position for the servomotors.
3. The control software reads the input device to obtain the user’s desired position.
4. The control software uses the desired position to update the position of the servomotors accordingly by writing to the PWM unit’s registers.

Essentially, our control software will be movement-pattern-agnostic with these changes as it no longer has any hard-coded pattern in its source code since the user provides the pattern at runtime.
Joysticks

The input device we will use in this lab the *PlayStation 2 joystick* shown in Figure 1. I sense nostalgia must already be building up!

![PlayStation 2 Joystick](image)

**Figure 1. PlayStation 2 Joystick**

The device consists simply of an input-controlled resistor on each axis, where the resistance ranges from approximately 30 Ω to 3.6 kΩ depending on the position of the joystick. The device has a 5-pin connector with the following connections:

- GND
- +5V
- VRx: Analog X-axis
- VRy: Analog Y-axis
- nSW: Digital button active-low

Analog to Digital Conversion

Hardware

Since we want to connect the joystick to our FPGA’s digital GPIO pins, we need to convert the analog output of the VRx and VRy pins to a digital signal before we can do anything with it. We will use an Analog to Digital Converter (ADC) for this purpose.

The extension board has 2 joysticks, each with 2 analog outputs, so we will need a 4-channel ADC. The extension board used in this course uses the **MCP3204**, a 4-channel, 12-bit ADC. Figure 2 shows the MCP3204’s package, and Figure 3 shows how the joysticks are connected to it on the extension board.

![MCP3204 Package](image)

**Figure 2. MCP3204**
Communication protocol

The ADC interfaces with the analog outputs of the joysticks and converts the analog value to a digital one. What remains to be determined is how this digital value is transmitted to the FPGA. This problem occurs quite often in computer systems and manufacturers have, over time, proposed standard communication protocols for this purpose. There are numerous communications protocols out there (SPI, I²C, 1-Wire, UART, …), and every device provides one or more such interfaces in order to be compatible with as many SoCs as possible.

The communication interface available on the MCP3204 is the Serial Peripheral Interface (SPI) bus. Figure 4 shows the details of the communication protocol between the ADC (SPI slave) and the FPGA (SPI master)\(^1\).

\(^{1}\) Note that the terms “master” and “slave” come up often in this field and are not Avalon-specific.
In the specific case of the MCP304, the same SPI bus is used both for commands and data. To obtain the digital value corresponding to one of the analog channels, you need to send a sequence of commands to the device, then the device responds with the corresponding data. Table 1 shows the configuration bits of the command sequence that needs to be sent to the MCP3204.

<table>
<thead>
<tr>
<th>Control Bit Selections</th>
<th>Input Configuration</th>
<th>Channel Selection</th>
</tr>
</thead>
<tbody>
<tr>
<td>Single/Diff</td>
<td>D2*</td>
<td>D1</td>
</tr>
<tr>
<td>1</td>
<td>X</td>
<td>0</td>
</tr>
<tr>
<td>1</td>
<td>X</td>
<td>0</td>
</tr>
<tr>
<td>1</td>
<td>X</td>
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<td>X</td>
<td>1</td>
</tr>
<tr>
<td>0</td>
<td>X</td>
<td>1</td>
</tr>
</tbody>
</table>

* D2 is a “don’t care” for MCP3204

**TABLE 1. CONFIGURATION BITS FOR THE MCP3204**

**FPGA-ADC Interface**

**Top-level Design**

In order to communicate with the ADC, we need to design a VHDL module that can supply commands to the ADC and that can read the converted values back through the SPI bus. We implement the *Avalon-MM slave* module shown in Figure 5 for this purpose.

![Figure 5. MCP3204 Top-Level Avalon-MM Slave Block Diagram](image)

The module consists of 2 parts:

- **MCP3204 Manager**: this unit interfaces with the host processor through an Avalon-MM slave interface and forwards requests to a custom SPI controller which then communicates with the MCP3204 and returns the conversion results. Table 2 shows the details of the unit’s register map. Note that all the unit’s registers are *read-only*, as it is impossible to “write” to the joysticks, so we can omit any write-related signals from the Avalon bus to simplify the design. The unit is supplied with a 50 MHz clock.
<table>
<thead>
<tr>
<th>Byte offset (from base)</th>
<th>Register Number</th>
<th>Name</th>
<th>Access</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>0x0</td>
<td>0</td>
<td>CHANNEL_0</td>
<td>RO</td>
<td>12-bit digital value of channel 0</td>
</tr>
<tr>
<td>0x4</td>
<td>1</td>
<td>CHANNEL_1</td>
<td>RO</td>
<td>12-bit digital value of channel 1</td>
</tr>
<tr>
<td>0x8</td>
<td>2</td>
<td>CHANNEL_2</td>
<td>RO</td>
<td>12-bit digital value of channel 2</td>
</tr>
<tr>
<td>0xC</td>
<td>3</td>
<td>CHANNEL_3</td>
<td>RO</td>
<td>12-bit digital value of channel 3</td>
</tr>
</tbody>
</table>

**Table 2. MCP3204 Manager register map**

- **MCP3204 SPI Controller**: this unit is responsible for performing the actual SPI communication to retrieve the converted values from the ADC. The unit is supplied with a 50 MHz clock, but all SPI communication must be done at 1 MHz.

Note that conversions are not done when a request arrives through the Avalon-MM slave interface, but are rather done continuously in the background: The Manager contains a state machine where it constantly instructs the SPI controller to perform a conversion when it is idle. The result of each conversion is stored in one of the Manager’s 4 internal registers and incoming requests on the Avalon-MM slave interface are served directly from the contents of these registers. This results in a more reactive system, as the master communicating with the MCP3204 Avalon-MM slave unit does not have to wait for a complete conversion cycle.

**SPI controller**

Your goal in this lab is to implement the SPI controller, so it is important to understand how it works. The SPI controller works in 2 conceptual phases:

1. The Manager instructs it which channel to convert.
2. The SPI controller communicates with the MCP3204 to obtain a converted value, then sends the received data back to the Manager.

Figure 6 explains the first phase of SPI controller’s operating procedure in more detail:

- The Manager waits as long as needed for the SPI Controller to be ready. The SPI Controller is considered unavailable if the busy signal is asserted.
- As soon as the SPI controller deasserts busy, it means that it is ready, so the Manager instructs it to start a new conversion. A new conversion is started by asserting the start pulse and supplying a 2-bit channel to the SPI controller. The SPI Controller captures the new instruction and becomes busy again since it starts converting the appropriate channel.

![Figure 6. SPI Controller — Phase 1](image-url)
Figure 7 explains the second phase of the SPI controller’s operating procedure.

- Once your SPI Controller has successfully converted a value, it passes the data to the Manager simply by asserting data_valid and putting the value on the data bus during one cycle.

**Figure 7. SPI Controller — Phase 2**

*Important:* Note that the start signal is a 50 MHz pulse, whereas the data_valid and data signals are 1 MHz pulses. You can see this due to the “clk” signal in Figure 6, whereas Figure 7 uses “SCLK”. We’ll see how to achieve something like this in the next section.

**Clock dividers**

The SPI controller is clocked at 50 MHz, but the SPI communication itself must be done at 1 MHz, so we need to figure out some way to slow down the clock frequency inside the SPI controller.

Students generally make the mistake of generating a slow “clock” by using logic inside the FPGA. For example, to generate a clock that runs 3x slower than clk_in, they would generate the clk_out waveform shown in Figure 8 and connect components as shown in the schematic below.

**Figure 8. Incorrect clock divider (3x)**

However, this is not the correct way to do things on FPGAs. Indeed, clock signals are routed through special channels to guarantee that the clock arrives at all components roughly at the same time (avoids clock skew). If you instead decide to “create” a clock manually by using logic within the FPGA, your clock will not be routed on the dedicated clock channels and will suffer from clock skew, therefore causing all logic driven by the custom clock to be unstable.
The correct way to generate slow clocks in FPGAs is to not generate a clock, but to instead generate a clock divider. A clock divider is a component that periodically generates a pulse which is then fed to the slower components. Figure 9 shows an example waveform and schematic of a correct divider that divides the clock frequency by 3.

![Figure 9. Correct clock divider (3x)](image)

Each component that requires a slow clock takes as input the standard clock of the FPGA (the one correctly routed through dedicated clock channels) along with the `enable_out` pulse generated by the clock divider. This pulse acts as an activation signal and triggers the operation of a component requiring a slow clock.

Note: We are telling you this, because you will be using a clock divider we provide you later on in the assignment, so it is important to understand how they work so you can use them correctly.

**ADC output**

The converted digital value you obtain as the output of the ADC depends on the position of the joystick when a conversion occurs. Figure 10 shows the values the ADC you should obtain when performing conversions.

![Figure 10. MCP3204 digital output depending on joystick position. The concrete behavior of the MCP3204 is shown on the left, whereas the expected behavior we want our API to exhibit is shown on the right.](image)
Practice
There is a lot to do in this lab, so let’s list the things so you can advance incrementally.

1. Implement the *hardware design* of the SPI controller in “hw/hdl/joysticks/hdl/mcp3204_spi.vh”.
   a. We gave you all the details you need to know in the assignment above, but you should still read section 5 of the *MCP3204 datasheet* to let everything sink in and to be sure you understand the protocol.
   b. To help you out, we provide the clock divider used for the SPI communication (and all associated signals). We suggest you read and make sense of the existing design before continuing.
   c. Draw the finite state machine (FSM) of your SPI controller on PAPER.
   d. Implement your FSM in the process named “STATE_LOGIC” in mcp3204_spi.vhd.
      i. **HINT 1:** Using ModelSim can truly be beneficial when implementing your design. You can use the stimulus generation file “tb_mcp3204_spi.vhd” we provide to generate instructions for the controller. Note that the file is not a testbench, but rather just generates instructions for the SPI controller, so you’ll have to compare the output of your design against Figure 4 to see if it is functioning correctly.
      ii. **HINT 2:** Pay attention to the fact that some signals need to be sent out on the *falling edge* of the SPI clock (SCLK), while others on the *rising edge* of SCLK. This can lead to difficult problems to debug! Use the “reg_rising_edge_sclk” and “reg_falling_edge_sclk” flags provided in mcp3204_spi.vhd to detect the correct edge!

2. Implement the *control software* for the MCP3204 Avalon-MM slave.
   a. Fill-in the following function in “sw/nios/application/joysticks/mcp3204/mcp3204.c”. This function is responsible for reading one of the 4 registers in the MCP3204’s Avalon-MM slave unit. **HINT:** It’s a 1-liner (or 2-liner to be safe 😊).

   ```c
   uint32_t mcp3204_read(mcp3204_dev *dev, uint32_t channel);
   ```

   b. Fill-in the following 4 functions in “sw/nios/application/joysticks/joysticks.c”. These are helper functions which wrap around the low-level “mcp3204.c” implementation and provide functionality at the joystick level of abstraction.

   ```c
   uint32_t joysticks_read_left_vertical(joysticks_dev *dev);
   uint32_t joysticks_read_left_horizontal(joysticks_dev *dev);
   uint32_t joysticks_read_right_vertical(joysticks_dev *dev);
   uint32_t joysticks_read_right_horizontal(joysticks_dev *dev);
   ```

   We want the values obtained from the 4 functions above to satisfy the API shown on the right diagram of Figure 10 (values *increase* from left to right, bottom to top). However, the physical readings obtained from the ADC are similar to the left diagram of Figure 10 (values *decrease* from bottom to top), so you'll need to find an algorithm (in software) to invert the reading of the Y axis for it to match the expected behavior of the system.
c. Fill-in the following function in “sw/nios/application/app.c”.

```c
uint32_t interpolate(uint32_t input,
        uint32_t input_lower_bound,
        uint32_t input_upper_bound,
        uint32_t output_lower_bound,
        uint32_t output_upper_bound);
```

This function is used to map the range of values obtained from the ADC readings of the joysticks to the range of values used by the PWM generator. This is a generic function which takes as arguments the upper and lower bounds of the input and output domains.

**HINT:** If you are skilled in the art of “Google-fu”, you will find implementations of this function on stackoverflow (under other names of course 😊), but we suggest you try to write it yourself.

**Conclusions**

At the end of this lab, you should have a system capable of controlling the direction of a servomotor based on the live position of a joystick. Congratulations for making it this far 😊. More cool stuff coming up in the next weeks.

![Figure 11. Expected Joystick & Servomotor Movement Behavior](image-url)